

Crocodile Optimizer-Based DC Chopper Control for Voltage Dip and Swell Mitigation in PMSG-Based WT

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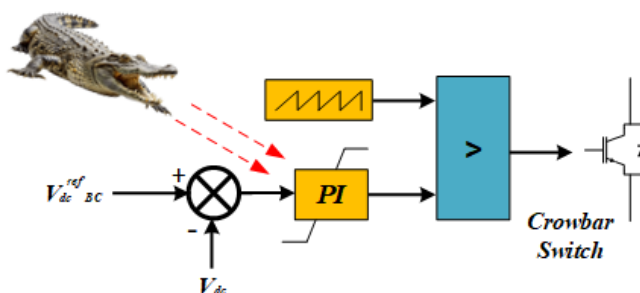
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ABSTRACT



The increasing penetration of wind energy systems requires advanced control strategies capable of maintaining stable operation during grid disturbances while complying with modern GC requirements. This paper proposes a COA-based control scheme for a DC chopper integrated into a PMSG system to enhance its FRTC under severe voltage disturbances. The COA is employed to optimally tune the controller parameters, ensuring effective regulation of the DC-link voltage and improved transient performance. The proposed approach is evaluated under four critical grid conditions, including three voltage dip scenarios corresponding to 100%, 80%, and 40% retained voltage levels, as well as a 20% voltage swell condition. Simulation results demonstrate that the proposed controller maintains the DC-link voltage close to its reference value of 1150 V, preventing excessive overvoltage during fault events and reducing stress on power electronic converters. Moreover, the control strategy satisfies GC requirements by providing appropriate reactive power support during VDs while ensuring controlled active power transfer. The optimized controller effectively suppresses electromagnetic torque oscillations, limits transient current peaks in the GSC, and enables rapid recovery of generator speed following fault clearance. Comprehensive MATLAB/Simulink studies confirm that the proposed COA-based DC chopper control significantly improves system transient stability, enhances grid-support capability, and ensures reliable operation under both voltage dip and swell conditions. In addition, the improved DC-link voltage regulation contributes to increased converter lifetime and reduced operational downtime, demonstrating the practicality and effectiveness of the proposed solution for modern wind energy conversion systems.

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1. INTRODUCTION

The current global energy demand exceeds the existing supply capacity, highlighting the urgent need for efficient and economically viable solutions to address future energy requirements [1]–[3]. In this context, prioritizing the integration of SSs has become essential in the development of modern energy systems [4][5]. Nevertheless, the global contribution of SSs to primary energy production remains inadequate to satisfy the increasing worldwide demand for energy and electricity [6][7]. Numerous developing nations are making substantial efforts to advance their energy sectors; however, achieving these objectives continues to present considerable challenges [8][9]. Energy resources are required to satisfy a wide range of human demands, including transportation, heating, and electricity supply for urban infrastructures. At the same time, assessing the long-term sustainability of these resources is of paramount importance. Sustainable sources ensure a continuous energy supply while significantly contributing to the reduction of pollution and greenhouse gas emissions, which are among the primary drivers of climate change [10][11]. SSs encompass naturally available resources such as biomass, geothermal energy, solar radiation, hydropower, wind energy, and ocean resources, all of which can be exploited to produce clean and usable energy. According to [12], each sustainable source possesses distinct advantages and limitations, which were systematically summarized in tabular form. To promote a sustainable and prosperous future, the United Nations introduced the SDGs in 2015, aiming for their comprehensive implementation by 2030 [13]. The framework of the SDGs also facilitates the monitoring and evaluation of progress toward achieving the objectives of the 2030 Agenda [14].

With respect to the SDGs, the expansion of SSs facilitates energy security across multiple sectors [15]. In addition to supporting environmental conservation and community development initiatives, SSs contribute significantly to meeting increasing energy demands. They have been recognized as an effective approach for addressing the energy challenges of the twenty-first century while simultaneously reducing negative environmental and climatic impacts [16][17]. Among renewable energy resources, wind energy provides several advantages owing to its technological maturity, reliable infrastructure, and economic feasibility [18]. Consequently, wind energy is expected to play an increasingly important role in the global energy supply in the years ahead [19][20].

With the increasing penetration of WFs into interconnected power systems, numerous countries have established specific interconnection standards and operational requirements for integrating wind farms into electrical grids, commonly referred to as GCs [21]. These GCs mandate that WFs remain connected to the power system during fault conditions and supply an adequate amount of reactive power (Q) at the PCC during VDIs [22][23]. Following fault clearance, wind farms are also required to restore and deliver active power (P) in compliance with GC specifications. For instance, the Danish GC stipulates that wind farms with a rated capacity greater than 11 kW must maintain grid connectivity for up to 0.5 s during a PCC voltage dip of as much as 80%, as illustrated in Area B of (Figure 1(a)) [24]. During this fault duration, the WF is required to inject Q to support the PCC voltage, as shown in (Figure 1(b)) [25]. The capability of WFs to satisfy GC requirements depends on both the generator technology and the adopted grid integration approach. Among the currently deployed wind energy conversion systems, the PMSG has gained considerable attention due to its favorable operational characteristics. In most applications, the PMSG is connected to the grid through full-scale BTBCs [26]. This configuration provides several benefits, including enhanced reactive power control capability, improved MPP tracking, and superior grid-side power quality performance.

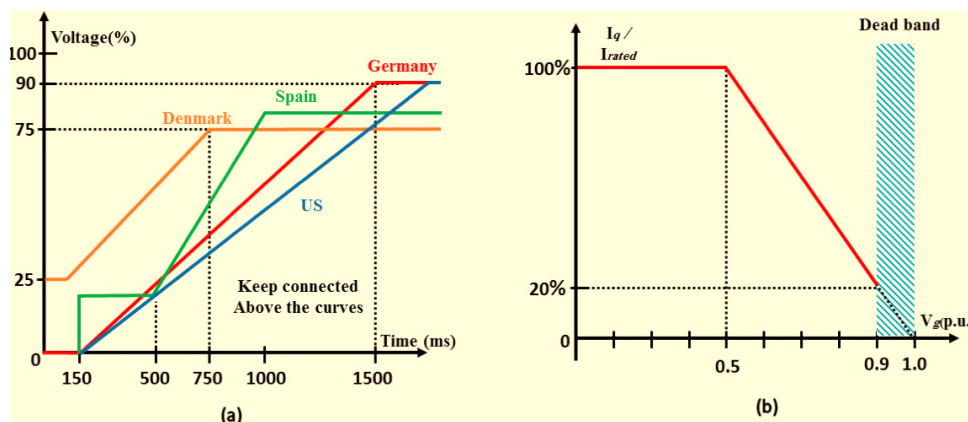


Figure 1. GC necessities that consist of: (a) a ride-through curve; and (b) a support curve of I_q

During grid faults, the wind farm generator continues producing electrical power; however, the amount of power delivered to the grid decreases because of the voltage dip at the PCC and the action of the GSC current limiter [27]. As a result, the mismatch between the input and output power of the converter during grid voltage dips leads to an over-voltage condition in the DCL [28]. To enhance and ensure the FRTC of WFs, various control and protection strategies have been proposed.

These approaches can generally be classified into two primary categories: (a) the use of external auxiliary equipment, and (b) the enhancement of BTBCs and pitch angle control systems. Each approach possesses distinct advantages and limitations, which have been comprehensively discussed in [29]. A recent control strategy involving the exchange of control functions between the GSC and MSC was proposed in [27]. Although this technique eliminates the need for additional external devices to achieve FRTC and demonstrates satisfactory performance under various VD conditions, the study presented in [30] indicated that such strategies may lead to unsatisfactory operational performance in certain cases. Despite the preference for approach (b) due to its low implementation cost and operational simplicity, it requires modifications to the pitch angle control mechanisms and converter control structures. Consequently, approach (a) is widely considered a reliable and GC-compliant solution for modern WFs.

Installing a BC across the DCL of the BTBC is considered one of the simplest and most cost-effective solutions for enhancing FRTC [31]–[33]. In contrast to other approaches categorized under method (a), such as the deployment of FACTS and ESS, which are generally characterized by high complexity and implementation cost [34], the BC-based approach offers a practical and economical alternative. One of the major advantages of the BC is its capability to operate as a temporary and rapid energy dissipation unit during fault conditions, making it suitable for application in various types of wind farms. Moreover, several studies have investigated system performance with the integration of BCs, as reported in [35]–[38]. In addition to wind farm applications, BCs have also been employed in AFDs and in FRT enhancement strategies for photovoltaic systems. A comprehensive overview of BC engineering applications in both (a) WFs and (b) AFD systems was presented in [39]. Various control techniques have been proposed for BC operation and management. Section 3 of this study presents a detailed discussion of several BC control methods, together with their respective advantages and limitations.

The existing literature presents various control and management strategies for PMSWGs. For instance, during GS disturbances, the study in [40] employed a Lyapunov-based MPC approach for the regulation of BTBCs. The proposed method demonstrated an improved transient response in DCL voltage (DCLV) regulation when compared with the conventional PI controller. However, the absence of converter current limiting mechanisms makes it difficult to accurately assess system performance under severe VD conditions. In [41], an adaptive fuzzy RST control technique was implemented for the regulation of both the MSC and the GSC. Additionally, the effectiveness of employing a high-selectivity filter under asymmetrical fault conditions was investigated.

Sliding Mode Control (SMC) is considered one of the most effective control strategies for enhancing system robustness and resilience [42]. This reliable control approach enables the GSC to satisfy GC requirements by injecting Q into the power system (PS), while simultaneously maintaining the DC-link voltage (DCLV) at its nominal level and ensuring appropriate system responses under both normal and fault operating conditions. Accordingly, SMC has been widely applied for the control of wind farms during fault scenarios. In [43], a configuration consisting of two PMSWGs connected to a common DCL was proposed, where a DC–AC converter interfaces the shared DCL with the power system. In this setup, the SMC technique was employed in the design and implementation of the control systems for all converters to achieve effective FRT performance. Despite its advantages, one of the main limitations of the SMC approach is the occurrence of chattering in the voltage and current signals of the converters. To address this issue, advanced techniques such as the Super-Twisting SMC (STSMC) method and the ERL approach have been introduced [44].

The proposed controller overcomes the limitations associated with existing BC control techniques, as discussed in Section 4. It should be noted that the proposed BC regulator can be applied to any BC integrated within the DCL of a BTBC, making it fully compatible with various BTBC control strategies. Furthermore, the proposed architecture is not restricted to the GC requirements of a specific country, as it can be adapted to comply with different international GC standards. In this study, the Danish GC is adopted as the reference framework for evaluation and analysis purposes.

The remainder of this paper is organized as follows. Section 2 presents the materials and methods employed in this study, including a brief description of the PMSWG model and system configuration, a review of existing BC control techniques, and the design and implementation of the proposed reliable BC controller

incorporating the COA approach. Section 3 discusses the simulation results and corresponding performance analysis. Finally, the main conclusions and findings of the study are summarized in Section 4.

2. MATERIALS AND METHODS

2.1. PMSWG Description & Modeling

The PMSWG model is divided into four key components (Figure 2): (1) mechanical section; (2) equivalent electric circuit in SRFMSC; (3) BTBC, DCL, BC, and the machine equations, in the SRFMSC, are shown in (1)–(6). The next equation's symbols are fully described in [45][46].

$$P_m = 0.5\rho A_{tur} C_p(\lambda, \beta) v_w^3 \quad (1)$$

$$\frac{d\omega_m}{dt} = \frac{1}{J_{eq}} (T_{Tur} - T_e - B_{eq}\omega_m) \quad (2)$$

$$T_{Tur} = 0.5\rho A_{tur} \frac{C_p(\lambda, \beta)}{\lambda} v_w^2 \quad (3)$$

$$\begin{cases} V_{ds} = R_s i_{ds} + L_s \frac{di_{ds}}{dt} - \omega_e L_s i_{qs} \\ V_{qs} = R_s i_{qs} + L_s \frac{di_{qs}}{dt} + \omega_e L_s i_{ds} + \omega_e \Phi \end{cases} \quad (4)$$

$$P_{gen} = \frac{3}{2} p \Phi |i_{qs}| \omega_m \quad (5)$$

$$\tau_e = \frac{3}{2} \frac{p}{\omega_m} \Phi i_{qs} \quad (6)$$

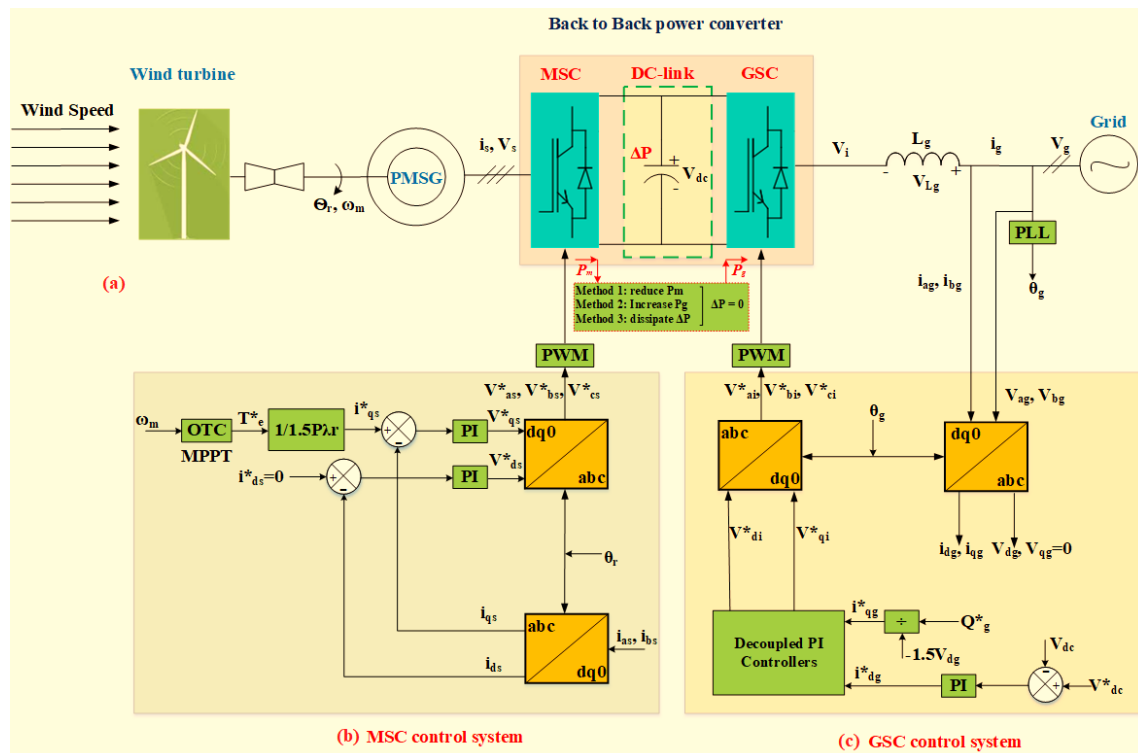


Figure 2. Addressed system

The pertinent equations for the GS electrical components and the BTBC-DCL are provided in (7) and (8). The BC-P loss flexibility only happens in (7) when there is a fault. Notably, the GS (+\sequence) electrical parts have been modeled using the well-liked SRFGSC approach [27].

$$\begin{cases} \frac{d}{dt} \left(\frac{CV_{dc}^2}{2} \right) = P_{gen} - P_{grid} & (\text{under normal conditions}) \\ \frac{d}{dt} \left(\frac{CV_{dc}^2}{2} \right) = P_{gen} - P_{grid} - P_{BC} & (\text{under grid faults}) \end{cases} \quad (7)$$

$$\begin{cases} E_{dq}^+ = R_f I_{dq}^+ + L_f \frac{dI_{dq}^+}{dt} + j\omega_f L_f I_{dq}^+ + V_{dq}^+ \\ E_{dq}^- = R_f I_{dq}^- + L_f \frac{dI_{dq}^-}{dt} - j\omega_f L_f I_{dq}^- + V_{dq}^- \end{cases} \quad (8)$$

2.1.1. Control Scheme of MSC

As illustrated in Figure 2(b), the MSC controller within the Synchronous Reference Frame MSC (SRFMS) consists of two control loops operating along the dq-axes. The primary objectives of the MSC control strategy are MPPT and generator power factor regulation. To minimize power losses and avoid demagnetization of the PMSWG, the reference current for the d-axis (I_d) is set to zero [24]. In addition, the OTC method is employed to achieve MPPT by generating the required q-axis current reference [35].

2.1.2. Control Scheme of GSC

To regulate (+\sequence) currents, a dual current controller is implemented in the (SRF_{GSC}), as illustrated in (Figure 2(c)). Owing to the high effectiveness of the (SRF-PLL) and the separation (+\sequence) (I/V) components in improving the performance of the GSC control system, the sequence separation method proposed in [24] is adopted in this study. This control strategy effectively suppresses the second-order harmonic components in the DCLV and enhances the active power (P) delivered to the PS during fault conditions. In addition to regulating the DCLV, the controller also ensures appropriate Q support to the grid. Furthermore, the reference values $P0^{ref}$ and $Q0^{ref}$ are determined according to the methodology presented in [27]. The desired (+\sequence) I_{dq} current references are calculated using (9) [47]. In each operating condition, $P0^{ref}$ is appropriately adjusted to achieve improved regulation of the DCLV.

$$\begin{bmatrix} I_{d-ref}^+ \\ I_{q-ref}^+ \\ I_{d-ref}^- \\ I_{q-ref}^- \end{bmatrix} = \frac{2}{3} \begin{bmatrix} V_{df}^+ & V_{qf}^+ & V_{df}^- & V_{qf}^- \\ V_{qf}^+ & -V_{df}^+ & V_{qf}^- & -V_{df}^- \\ V_{qf}^- & -V_{df}^- & -V_{qf}^+ & V_{df}^+ \\ V_{df}^- & V_{qf}^- & V_{df}^+ & V_{qf}^+ \end{bmatrix}^{-1} \begin{bmatrix} P_0^{ref} \\ Q_0^{ref} \\ 0 \\ 0 \end{bmatrix} \quad (9)$$

2.2. Review of BC Control Methods

The BC primarily consists of two main components: a DC chopper switch and a resistor, as illustrated in (Figure 2(a)). When the DCLV exceeds a predefined threshold, the resistor is connected into the circuit to dissipate the excess energy stored in the DC-link, thereby maintaining the voltage within permissible operating limits. Several control strategies have been proposed for BC regulation and duty cycle (D) determination. In [41], the deviation between the measured DCLV and its reference threshold is processed through a PI controller to determine the modulation index (m), which is subsequently used to calculate the duty cycle, as shown in (Figure 3(a)) [48]. In contrast, the approach presented in [49] directly compares the DCLV with its threshold value, and the resulting signal is utilized to generate the gate pulse for the BC switch, as illustrated in (Figure 3(b)) [48]. The primary objective of these controllers is to maintain the DCLV at a constant value. Consequently, the capability of the GSC to provide Q support to the grid has not been fully utilized. Furthermore, these control methods are not sufficiently suitable for operation during VD conditions, where mitigation of system imbalance caused by the VDs is required.

Reference [50] proposed a control strategy in which the BC is activated based on the current phasor (CP) signal of the GSC. According to this approach, during a VD, the GSC attempts to maintain the active power

(P) delivered to the PS at a constant level. As a result, the occurrence of a VD leads to a significant increase in the magnitude of the current phasor injected into the PS. The authors also indicated that one of the limitations of this technique is the potential oscillatory behavior of the regulator. In addition, several other drawbacks may arise, including: (1) reduced operational effectiveness under varying wind speed conditions, which alter the current phasor supplied to the PS, and (2) the omission of a current limiter block within the converter control structure. Considering these limitations, further investigation is required to accurately assess the effectiveness and reliability of the proposed control strategy.

In [48], the modulation index (m) and, consequently, the D of the BC were determined, as illustrated in (Figure 3(c)) by evaluating the power fluctuations among the PMSWG, the grid, and the ESS, relative to the comparatively small BC power. In these studies, the ESS was connected across the DCL. However, the main limitations of this approach arise from delays associated either with BC activation or with the regulator response. Due to the absence of effective voltage regulation, the DCLV may reach excessively high amplitudes, preventing the system from maintaining the desired voltage level. Consequently, the regulator is primarily limited to balancing the input and output P of the DCL, without providing adequate direct control of the DCLV.

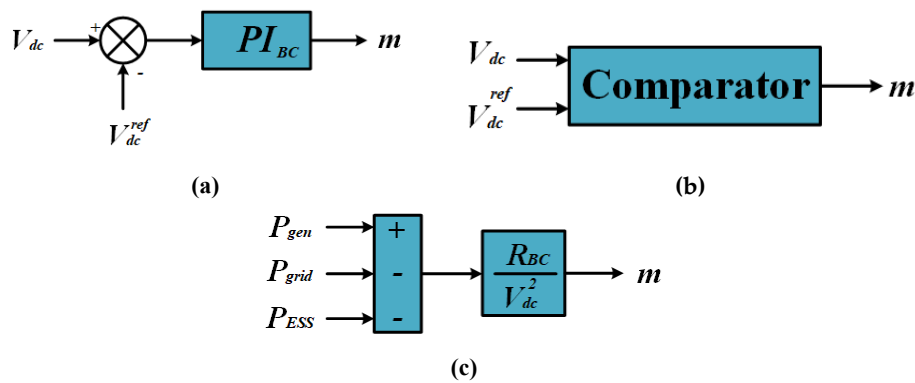


Figure 3. BC control structure

The output signal (m) is expressed in (10) as follows

$$m = \frac{P_{gen} - P_{grid} - P_{ESS}}{\frac{V_{dc}^2}{R_{BC}}} \quad (10)$$

In several studies, such as [51], the BC approach has been integrated with additional techniques to improve the performance of PMSWGs during FRT operation. Table 1 summarizes these approaches, which combine the BC system with pitch angle control, ESS, STATCOMs, and enhanced BTBC control schemes. The comparison among these methods is primarily based on two key performance indicators: the capability of injecting Q into the PS and the effectiveness of DCLV regulation. It can be observed that the issue of Q support to the grid has not been comprehensively addressed in the existing literature. Moreover, in many of the previously proposed approaches, including those discussed above, the integration of multiple auxiliary systems results in additional costs for WF operators and stakeholders. Therefore, enhancing the BC control strategy or integrating the BC approach with improved BTBC control methods represents a practical and cost-effective solution for improving FRTC in a highly cost-sensitive wind energy sector.

Table 1. A Summary of the Various Controls That PMSWG Uses in Conjunction with the BC Technique

The Ability of a Combination of Control Systems	Control System Mutual with BC			
	Pitch Regulator	ESS Regulator	STATCOM Regulator	BTBC Regulator
DCLV Control	[52]	[53]	[54]	[55]
Q Injection	[52]	[53]	[54]	[55]

In this study, the performance of the (PMSWG) is further enhanced through the adoption of a modified (GSCCS) proposed in [27][45]. In addition to the primary objective of investigating and evaluating an improved Braking Chopper (BC) regulator against conventional BC control approaches, the modified GSCCS contributes significantly to system performance enhancement. The proposed GSCCS modification determines the reference (P) of the (GSC) using (DCLV) information rather than generator speed measurements. Under

fault conditions, the activated GSCCS enables appropriate (Q) injection into the power system (PS) in compliance with WF GC requirements. Furthermore, this control strategy effectively mitigates voltage imbalance at the PCC by supplying asymmetrical current to the grid, as detailed in [27].

Furthermore, to provide a deeper understanding of the BC regulator and its operating principles, this study also examines the configuration of an industrial BC system. In most commercially available BC units designed for WF applications, six switches are employed for BC configuration, as illustrated in (Figure 4) [39]. The first three switches (S1S_1S1, S2S_2S2, and S3S_3S3) are configured according to predefined tables based on the magnitude of the instantaneous load to be dissipated and the thermal time constant (TTC) of the DC resistor (DCR). Switch S4S_4S4 is adjusted according to the operational characteristics and switching frequency of the DC chopper, whereas switch S5S_5S5 is configured in relation to the grid voltage level. In practical implementations, the BC settings must satisfy both the transient load requirements and the TTC constraints of the DCR. The final switch (S6S_6S6) is assigned a unique identifier for each BC unit and is primarily utilized for communication purposes. In addition, up to eight BC circuits can be integrated within a single converter system, as discussed in [39]. An important observation is that industrial BC systems currently employed in practice mainly rely on DCLV-based control strategies. Consequently, they exhibit limitations similar to those associated with conventional DCLV-controlled BC methods. A comprehensive analysis of BC thermal capacity characteristics is provided in [39].

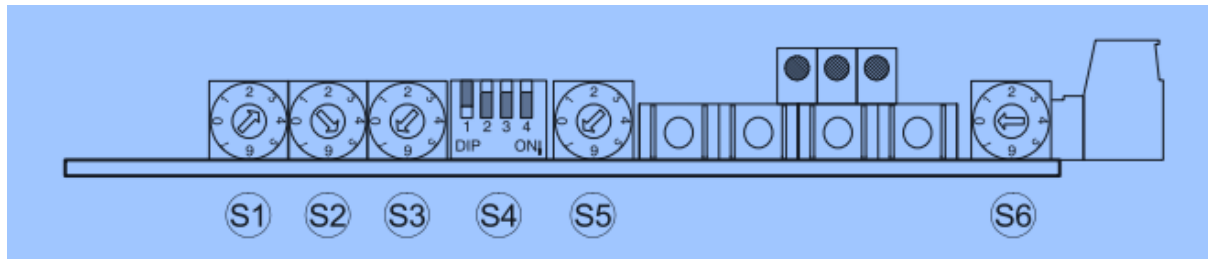


Figure 4. Archetypal BC situation switches cast-off via the DCL

2.3. Proposed BC control system

The proposed control system utilizes the COA to facilitate the safe operation of a PMSWG during grid faults. In this framework, the COA is employed to optimize the performance of an industrial BC installed across the DCL. The primary objective of the COA-based controller is to maintain the DCLV within acceptable bounds by balancing the input-output power during transient grid disturbances, such as voltage dips and swells. By mimicking the strategic hunting behavior of crocodiles, the algorithm fine-tunes the controller gains to minimize oscillations in electromagnetic torque, rotor speed, and grid-side converter currents. This approach ensures the system meets GC requirements for FRTC while providing reliable grid support through reactive power injection.

2.3.1. COA Mathematical Model

The COA is a nature-inspired metaheuristic that mimics the ambush hunting mechanisms of crocodiles, specifically focusing on infiltration (exploration) and raiding (exploitation) as seen in Figure 5. The search agents represent potential solutions, organized in a population matrix X where n is the population size and D is the dimension of each agent. Figure 6 depicts the COA flowchart.

To guide the search process, a conversion factor (C_f) is utilized to balance exploration and exploitation:

$$C_f = C_0 \cdot \left(1 - \frac{1}{2T_{max}}\right) \quad (11)$$

where C_0 is set to 0.8; t is the current generation, and T_{max} is the maximum number of generations as in (11).

Step 1: Premeditation and Waiting Phase (Exploration)

When $C_f > 0.6$, the algorithm enters the exploration stage. Crocodiles swim randomly to exchange information and obtain high-quality solutions from the current best individual (X_{gbest}). The mathematical model for this position update is:

$$\begin{cases} Step = c_1 \cdot (X(t) - X_{r_1}(t)) + c_2 \cdot (X_{g_{best}} - X(t)) \\ X(t+1) = X(t) + Step \end{cases} \quad (12)$$

where c_1 and c_2 are random numbers in (0, 2) and r_1 is a random integer as seen in (12) and (13).

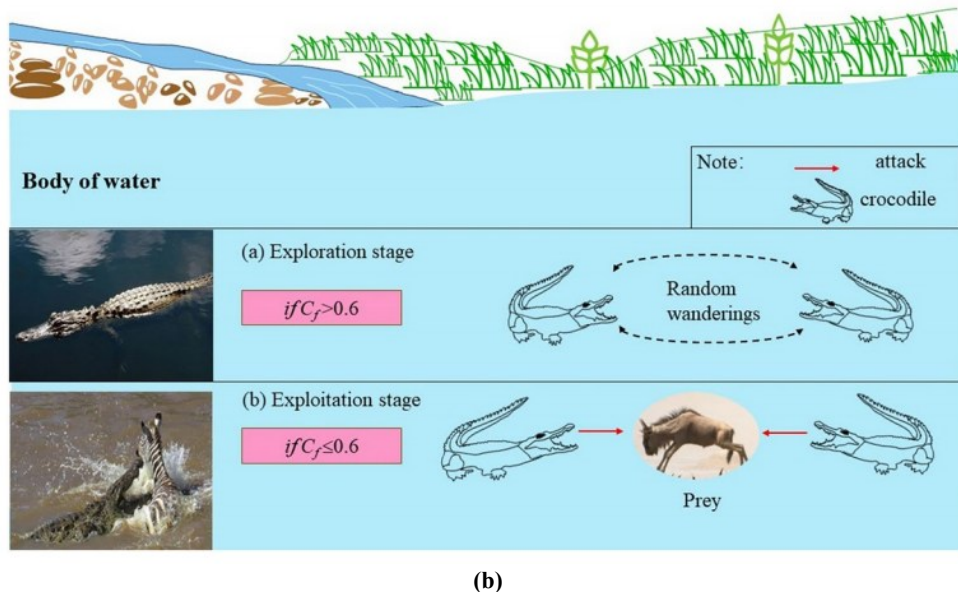
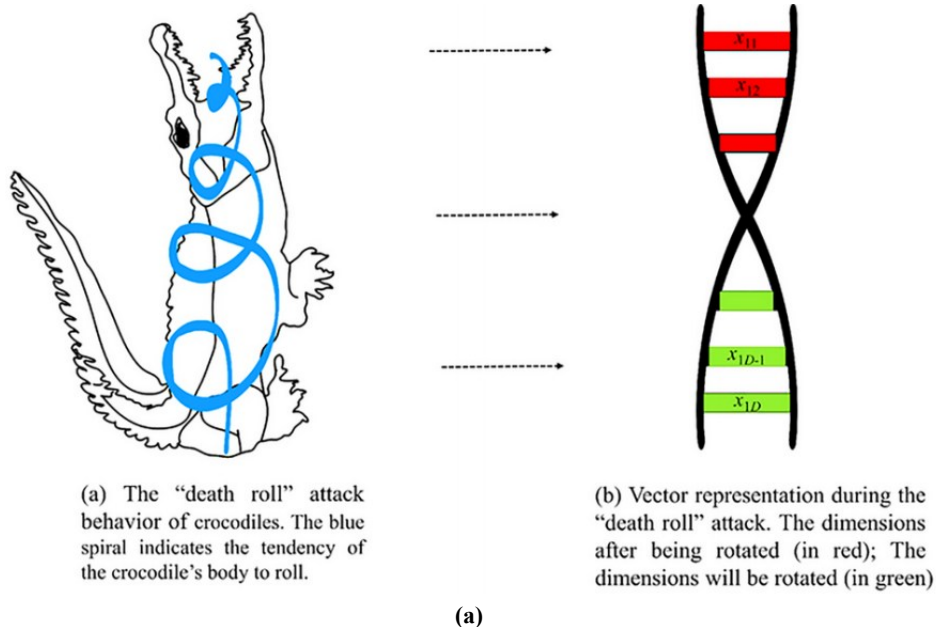


Figure 5. (a) COA strategy: The ‘Death Roll’ hunting behavior of crocodiles; **(b)** COA strategy: Schematic diagram of crocodile hunting activity areas

Step 2: Attacking, Hunting, and Diet Phase (Exploitation)

When $C_f \leq 0.6$, the algorithm performs exploitation, mimicking a **sneak attack** and the signature “**death roll**”.

- **Sneak Attack:** The position changes based on the proximity to the attack range, controlled by parameter C :

$$\begin{cases} X_a(t) = X_{pbest} + (X(t) + C) \cdot \cot(\theta - C), & \text{if } p > 0.5 \\ X_a(t) = X_{pbest} + (X(t) + C) \cdot \tan(\theta + C), & \text{else} \end{cases} \quad (13)$$

where p is a random number in $(0, 1)$ and θ is a time-varying angle.

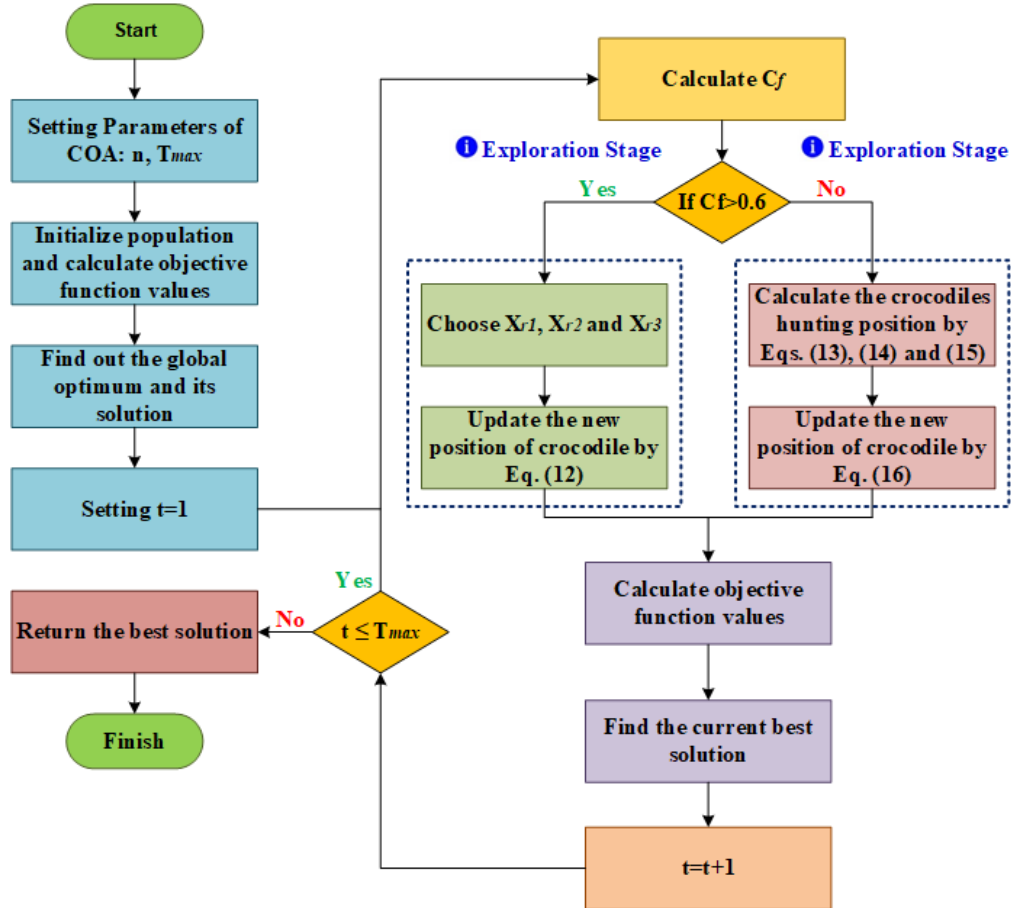


Figure 6. COA flowchart

- **Death Roll Attack:** This simulates the crocodile's rotation to tear prey, utilizing random integers r_2 and r_3 to influence dimension-wise updates:

$$\begin{cases} X_{d1} = (X(t) - X_{r2}(t)) \cdot \exp\left(-\frac{t}{T_{max}}\right) \cdot \cos\left(2\pi \cdot \frac{t}{T_{max}}\right) \\ X_{d2} = (X_{r2}(t) - X_{r3}(t)) \cdot \exp\left(-\frac{t}{T_{max}}\right) \cdot \cos\left(2\pi \cdot \frac{t}{T_{max}}\right) \end{cases} \quad (14)$$

The integrated hunting model is then defined as

$$X_h = \frac{1}{2}(X(t) - X_{d1}) + \frac{1}{2}(X_{r2} + X_{d2}) \quad (15)$$

- **Diet Stage:** After the attack, the final position is updated based on the prey's physical energy exertion (E):

$$\left\{ \begin{aligned} X(t+1) &= X_{gbest} - E \cdot X_h \\ E &= \frac{1}{2} \left(\exp\left(-\frac{t}{T_{max}} \cdot C\right) + \exp\left(-\frac{t}{T_{max}} \cdot |X_{gbest} - X_h|\right) \right) \end{aligned} \right\} \quad (16)$$

where E represents the diminishing stamina of the captured prey

2.3.2. Application of COA

COA is applied on the BC controller of PMSWG to fine-tune the PI controllers' gains, as seen in Figure 7, where the configuration of the DCL in the BC-BTBC is depicted in Figure 8. This tuning is performed to improve the dynamic performance of the investigated system during normal and abnormal conditions. The optimization of the systems under study (taking control cost (CC) into account) can be formally represented in (17). The objective function for CC that is employed is written as follows:

$$\text{Minimize } F(x) = \int_0^T W_1 |P - P^*| + W_2 |\omega_m - \omega_m^*| + W_3 |Q_s - Q_s^*| + W_4 |V_{dc} - V_{dc}^*| + W_5 |C_p - C_p^*| \quad (17)$$

where W_1, W_2, W_3, W_4 , and W_5 are constants used for the estimation of the CS function, which is 4×10^5 here. T denotes the average time, and 100 and 6 are the number of iterations and agents, respectively.

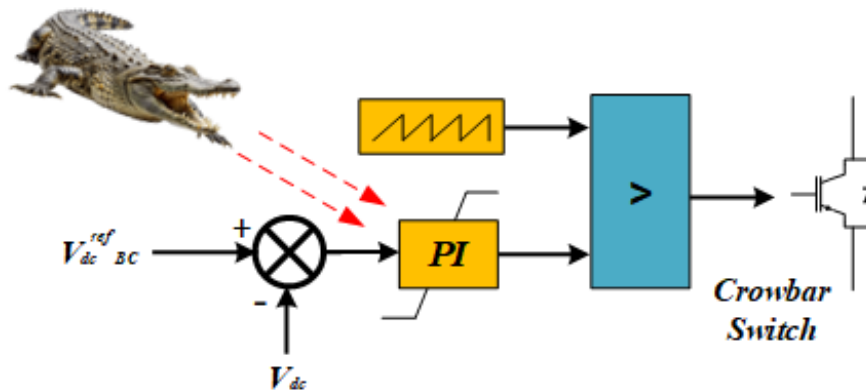


Figure 7. Control of the Crowbar system based on COA

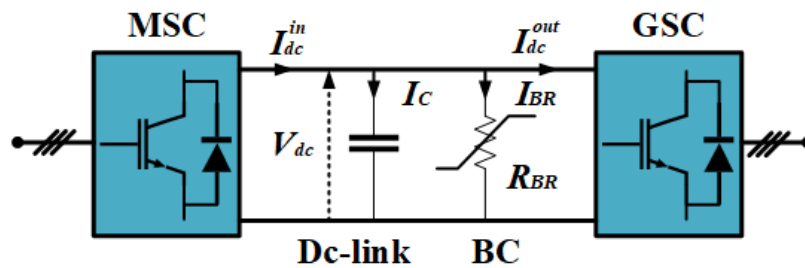


Figure 8. Configuration of the DCL in the BC-BTBC

3. RESULTS AND DISCUSSIONS

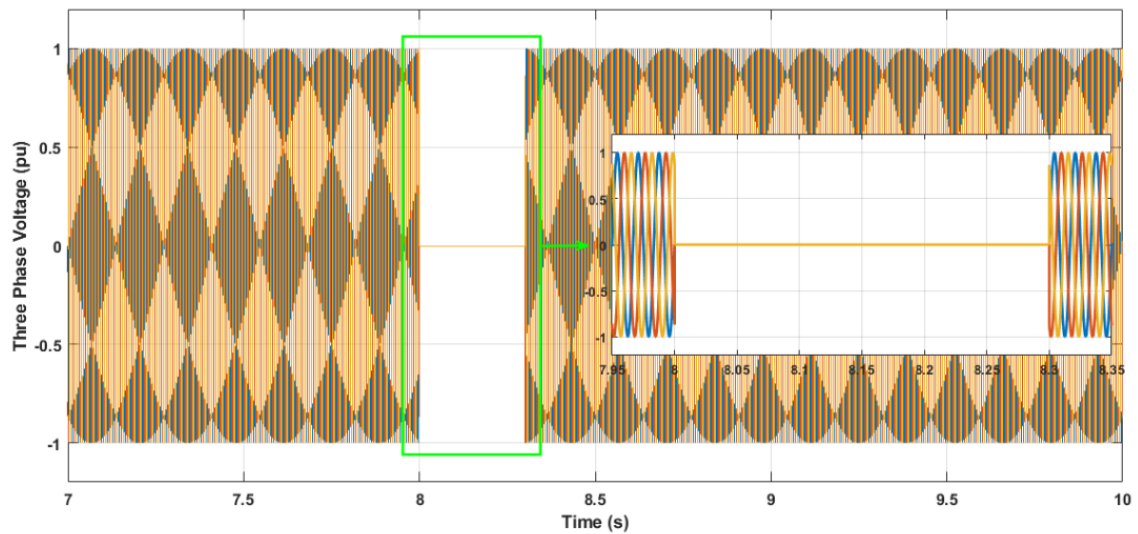
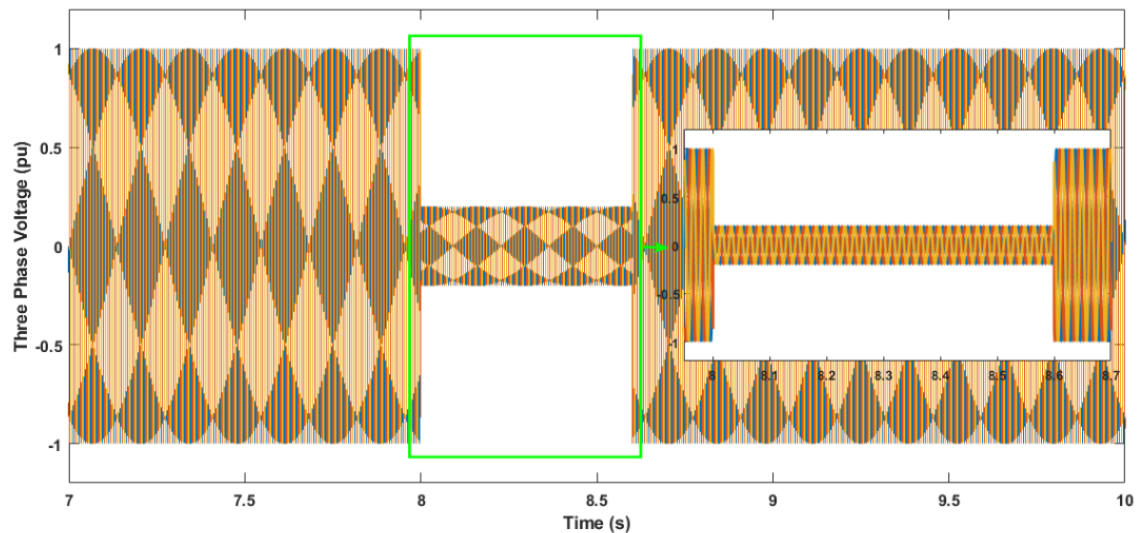
Under unbalanced grid voltage conditions, the primary challenge encountered by the system is the excessive rise in the DCLV. Therefore, maintaining the DCLV within acceptable operating limits serves as a key indicator of the effectiveness of the proposed control strategies. To evaluate the performance of the proposed FRTC enhancement techniques, four simulation scenarios were implemented using the MATLAB/Simulink environment. Considering the short duration of the transient disturbances, the wind speed was assumed to remain constant at 12 m/s throughout the fault period. The investigated scenarios were defined according to the remaining voltage level at the PCC and were denoted as follows: **S0** (100% voltage dip), **S0.2** (80% voltage dip), **S0.6** (40% voltage dip), and **S1.2** (20% voltage swell). To demonstrate the effectiveness of the COA, the transient response of the PMSWG parameters was analyzed using a COA-PI-controller-based BC under the aforementioned operating conditions. The investigated system variables include: V_{abc} , P_s , Q_s , I_{d-GSC} , I_{q-GSC} , T_{em} , V_{dc} , and ω_r . The studied PMSWG system was initially assumed to operate under steady-state (SS) conditions with a Q_s of ≈ 0 , a system voltage $V_{rms} = 1$ pu, and a $P_s = 0.96$ pu. The parameters of the investigated PMSWG and BC systems (Table 2) are presented in [35].

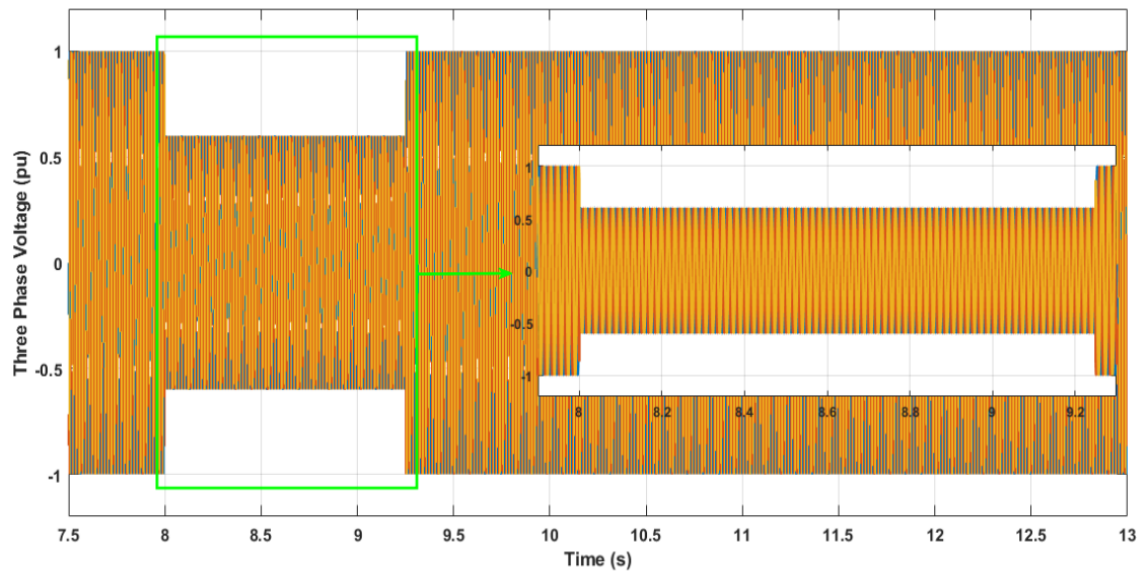
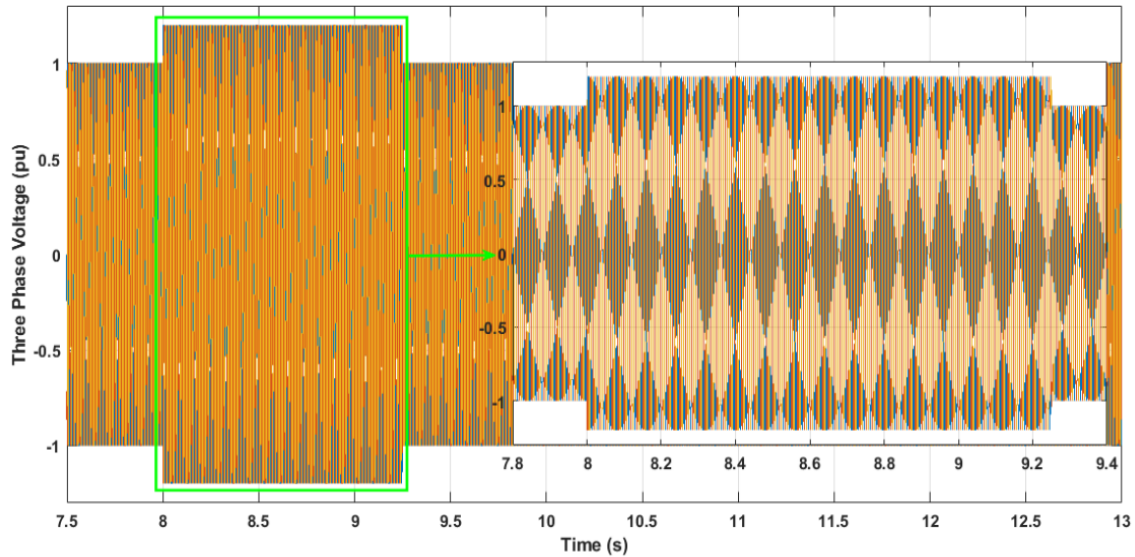
Table 2. Simulated Parameters under the Studied Cases

Parameters	Studied cases using COA			
	S0	S0.2	S0.6	S1.2
Overvoltage at V_{dc}	1.04 pu	1.026 pu	1.022	1.022 pu
P_S change	0.59 pu $\rightarrow$ 0.96 pu	0.34 pu $\rightarrow$ 0.957 pu	0.73 pu $\rightarrow$ 0.957 pu	0.9566 pu $\rightarrow$ 0.9583 pu
I_{d-gsc} ripple	0.2263 pu	0.21 pu	0.21 pu	- 0.1731 pu
I_{q-gsc} ripple	± 0.03 pu	± 0.02 pu	± 0.015 pu	0 pu
ω_r	1.09 pu	1.091 pu	1.091 pu	1.091 pu
T_{em}	0.8956 pu	0.8955 pu	0.895 pu	0.8942 pu

3.1. Three Phase Voltage Profiles (Vabc)

V_{abc-0} : Figure 9(a) shows a 100% voltage dip (S0) lasting for 300 ms, where the voltage drops to zero and recovers at $t=8.3$ s. $V_{abc-0.2}$: Figure 9(b) illustrates an 80% voltage dip (S0.2) lasting for 600 ms, with the voltage dropping to 0.2 pu and recovering at $t=8.6$ s. $V_{abc-0.6}$: Figure 9(c) displays a 40% voltage dip (S0.6) for 1.25 s, where the voltage drops to 0.6 pu and recovers at $t=9.25$ s. $V_{abc-1.2}$: Figure 9(d) shows a 20% voltage swell (S1.2) for 1.25 s, with the voltage rising to 1.2 pu before returning to normal at $t=9.25$ s.

(a) V_{abc-0} (b) $V_{abc-0.2}$

(c) $V_{abc-0.6}$ (d) $V_{abc-1.2}$ **Figure 9.** Investigated voltage profiles

3.2. Power and Torque Response

Active Power (P_S): Figure 10(a) displays the P_S (pu) delivered to the grid. During VDs (S0, S0.2, S0.6), there is a significant drop in P_S transmission at $t=8$ s, followed by a controlled recovery to steady state (approximately 0.96 pu) once the fault is cleared. Reactive Power (Q_S): Figure 10(b) shows the Q_S (pu). To comply with GCs and support the grid during dips, the system injects Q_S , peaking during the 100% dip (S0) and 80% dip (S0.2). Mechanical Torque (T_{em}): Figure 10(c) illustrates the Electromagnetic Torque (pu). The figure highlights transient fluctuations and ripples at the beginning and end of the faults, with the COA-based controller aiming to minimize these oscillations to ensure a smoother response.

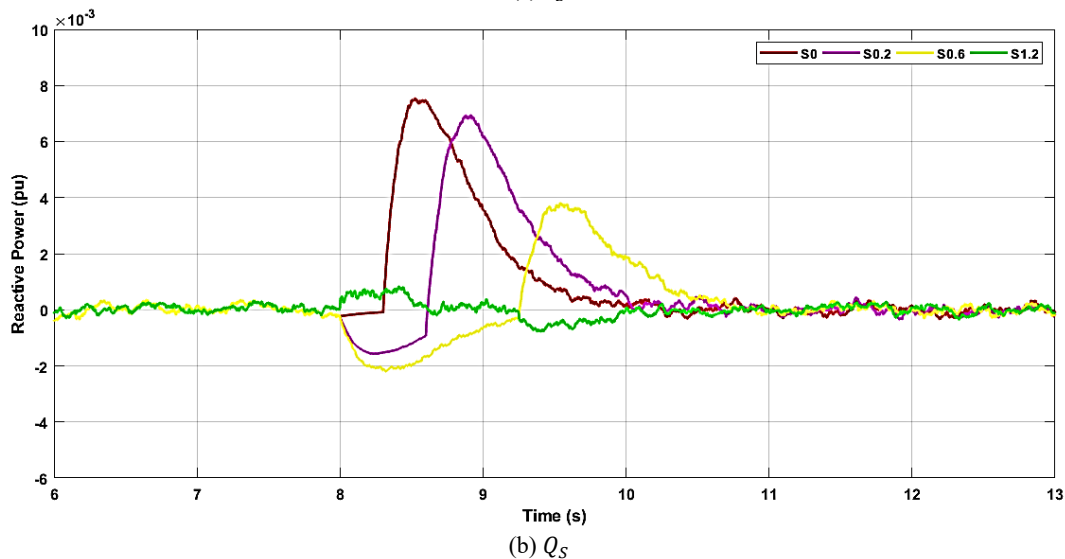
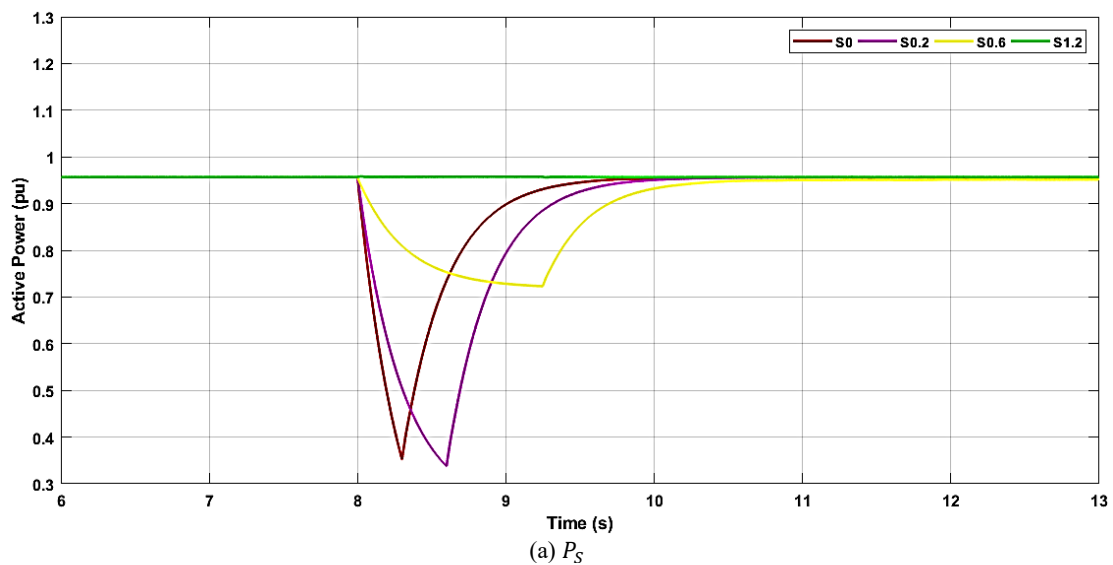
3.3. GSC Currents

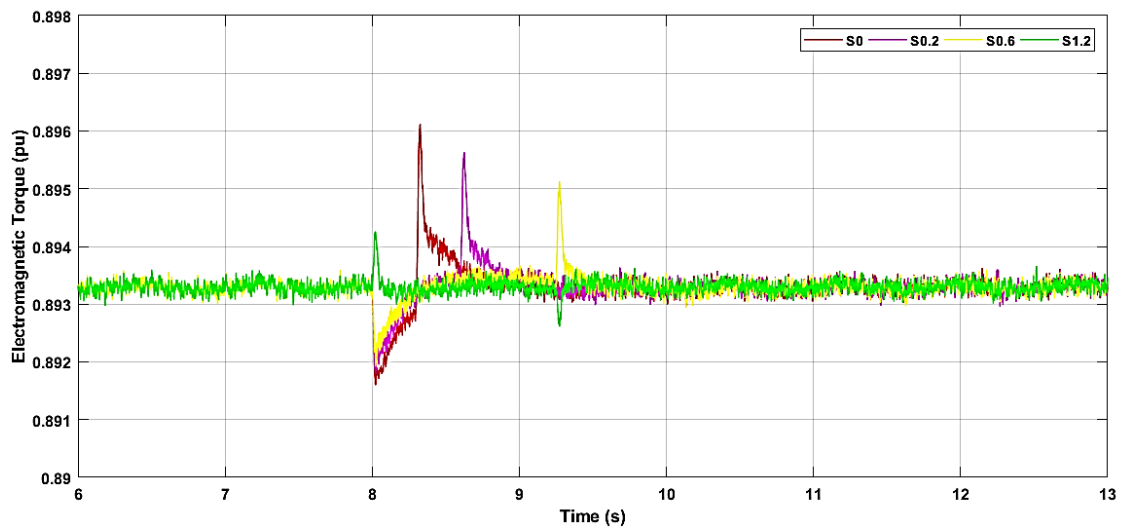
Direct Current (I_{d-GSC}): Figure 11(a) represents the i_d of the GSC. It shows the shifts in current required to manage active power and DCL stability during the fault periods. I_{q-GSC} : Figure 11(b) Shows the i_q of the GSC. This parameter experiences ripples and transient spikes at the fault inception ($t=8$ s) and clearance, particularly in the more severe 100% dip scenario.

3.4. System Stability Parameters

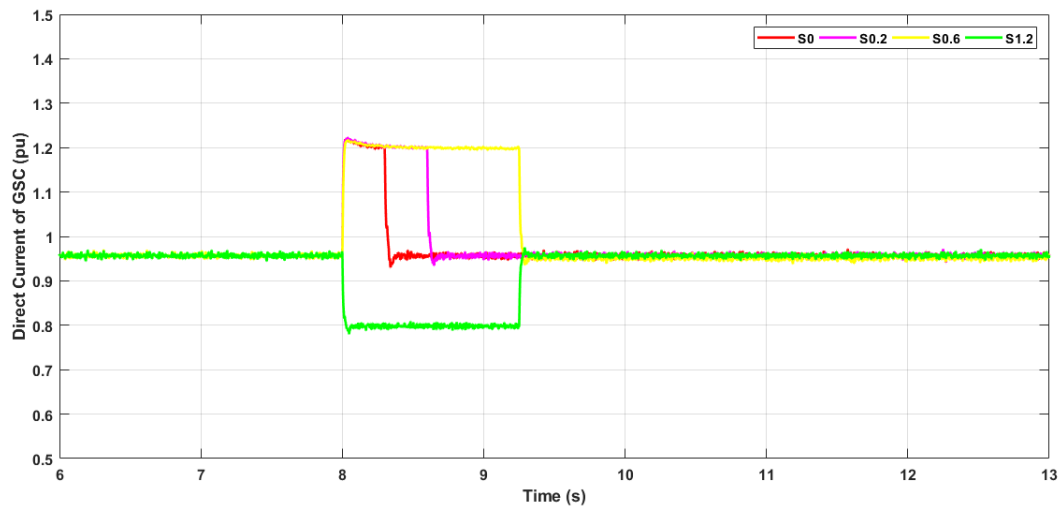
DCLV in Figure 12(a) displays the DCL capacitor voltage (V). This is a critical parameter for evaluating the COA-based controller. The figure shows that despite the severe faults, the controller successfully maintains the voltage near the reference value (1150 V), preventing dangerous over-voltages. Angular Speed (ω_r): Figure 12(b) Shows the angular speed (pu) of the generator. The rotor speed oscillates during the fault transients but is stabilized by the control system, returning to its steady-state value of approximately 1.09 pu.

Figure 13 displays the DCLV values for each of the scenarios that were looked at to assess the proposed method. A comparison with earlier published techniques based on FLC, MPC, SMC, or optimizers is also included in Table 3. Although the more advanced and contemporary types show respectable transient responsiveness, FCS-MPCs are vulnerable to the complexity and volatility of currents and DCLV following fault elimination due to their low switching frequency. The DCLV values did not match the required threshold (higher than 1.1 pu) in each of the VD scenarios. The recommended control methods not only enhanced the PMSG settings' abrupt response but also preserved the DCLV value fulfilled in every situation that was examined.

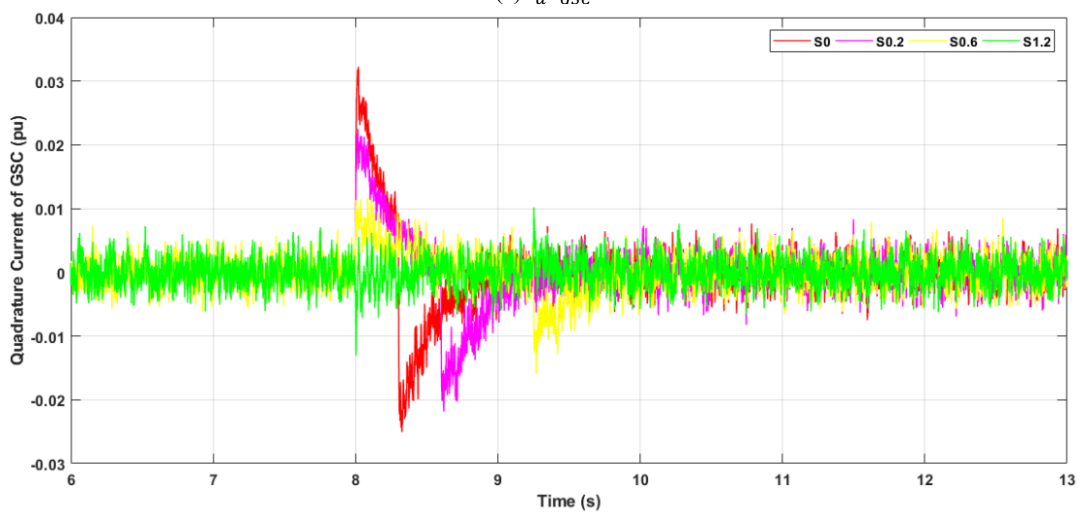




(c) T_{em}
Figure 10. Power and Torque



(a) I_{d-GSC}



(b) I_{q-GSC}

Figure 11. Transient SP as a result of 40% VD

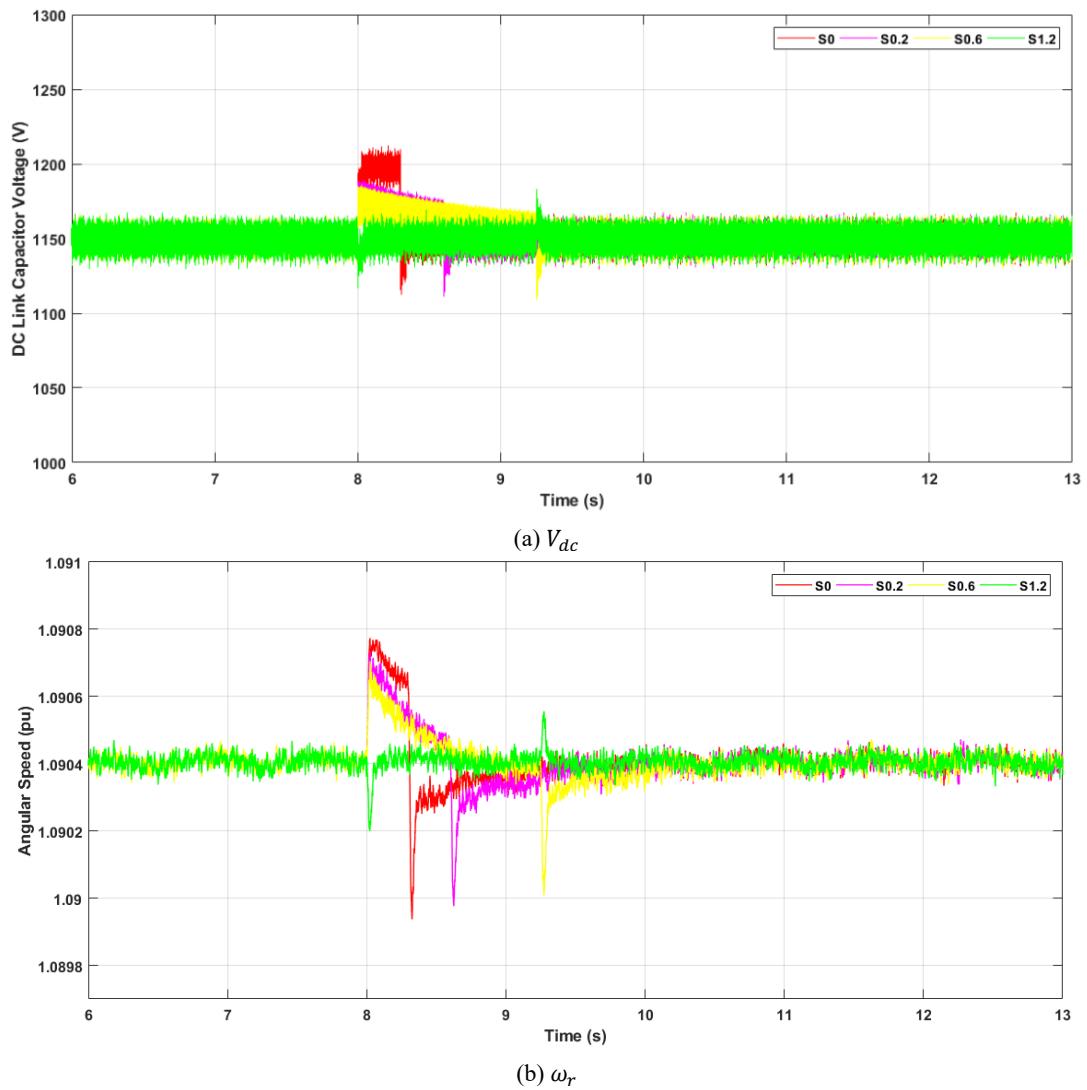


Figure 12. System Stability Parameters

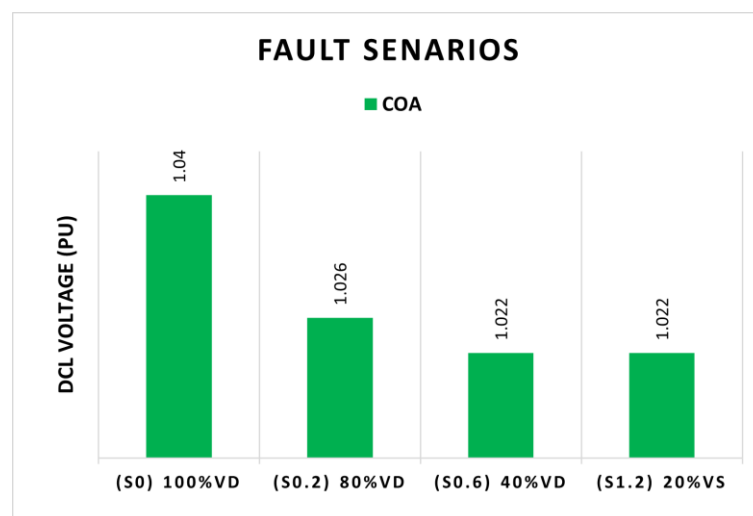


Figure 13. DCLV values under the investigated scenarios

Table 3. Comparative Analysis with Previously Published Techniques

Ref	Publisher	Year	Technique	Smooth Operation	Transient response	Complexity	DCLV during faults			
							DCLV (pu)	VD (pu)	DCLV (pu)	VS (pu)
[56]	MDPI	2023	NLESO and GFTSMC	√	VH	H	1.154	1	X	X
[35]	Taylor&Francis	2023	WHO-PI	√	VH	L	1.043	0.9	X	X
[57]	MDPI	2022	ESS and BC	√	VH	VH	1.01	0.4	X	X
[58]	MDPI	2023	ILP-PLL	√	VH	VH	1.004	0.35	X	X
[45]	SAGE	2021	FLC	X	H	H	1.047	0.9	X	X
[59]	Elsevier	2024	Modified MSC	√	H	L	1.4	asymmetrical grid faults	X	X
			COA-PI (proposed)	√	VH	L	1.05, 1.043, 1.035	1, 0.7, 0.3	1.022	0.2

4. CONCLUSIONS

The dynamic response of PMSWG under VD and VS scenarios was examined in this work employing a designed BC with COA. The system under investigation has been mathematically modeled, and COA has been applied to operate the system. Four scenarios are investigated. Table 4 summarizes all of the results that were collected. To demonstrate the efficiency and effectiveness of C, the system is run using several control schemes. The findings demonstrated that COA improves FRTC by maintaining the DC within allowable bounds, in addition to improving the dynamic response of the parameters. In comparison to A and B, the excess voltage at DCL decreased by 93.5% and 19.231%, respectively, with C in the worst case (100% VD). When it came to smoothing the power transmitted to the utility grid and reducing oscillations in the rotor's angular speed, generator electromagnetic torque, and direct and quadrature currents, C outperformed B and A. Additionally, to illustrate the contribution of the proposed strategy to sustainable development, a comparison study (Table 5) including more than a dozen published works from respectable publishers has been conducted. In conclusion, the suggested approach will contribute to the attainment of high wind energy penetration scenarios.

DECLARATION

Author Contribution

All authors contributed equally to the main contributor to this paper. All authors read and approved the final paper.

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Conflicts of Interest

The authors declare no conflict of interest.

ABBREVIATIONS

The following abbreviations are used in this manuscript.

AFD	:	adjustable frequency drives
BC	:	braking chopper
BR	:	braking resistance
BTBC	:	back-to-back converter
PI	:	proportional-integral
ERL	:	exponential reaching law
ESS	:	energy storage system

GC	:	grid-code
GSC	:	grid-side converter
KCL	:	Kirchhoff's current law
FRTC	:	Fault ride-through capability
MPPT	:	maximum power point tracking
FACTS	:	Flexible AC Transmission Systems
SDGs	:	Sustainable Development Goals
SS	:	Sustainable sources
WF	:	Wind farm
PMSG	:	permanent magnet synchronous generator
SMC	:	sliding mode control
SPWM	:	sinusoidal pulse width modulation
SRF	:	Synchronous reference frame
PLL	:	Phase-locked loop
VD	:	Voltage dip
MSC	:	machine-side converter
OTC	:	optimal torque control
PCC	:	point of common coupling
DCL	:	DC link

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