# Induction Motor Speed Control Using PID Tuned by Particle Swarm Optimization Under Vector Control

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# ARTICLE INFORMATION

## **ABSTRACT**

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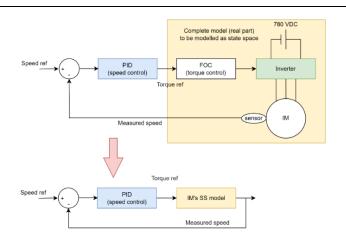
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Induction motors (IMs) are widely used in industrial applications due to their cost-effectiveness, durability, and low maintenance requirements. This study investigates the speed control of an induction motor using vector control combined with a PID controller whose parameters are tuned via Particle Swarm Optimization (PID-PSO). A reduced-order small-signal state-space model is derived from a detailed nonlinear IM model to facilitate efficient controller tuning while maintaining fidelity to real-world behavior. The PID parameters are optimized using PSO, with the Integral of Absolute Error (IAE) selected as the objective function due to its ability to penalize longduration deviations and reflect steady-state performance. The optimized PID controller is then validated on the full nonlinear IM model under speed and load variations. Simulation results demonstrate that PID-PSO significantly outperforms manually tuned PID control in terms of tracking accuracy, reducing IAE by 37.79% and 14.76% under speed and load variation conditions, respectively. However, this improvement comes at the cost of slightly slower settling time. These results highlight a trade-off between accuracy and transient response, motivating future research on multiobjective optimization to balance conflicting criteria such as robustness, energy efficiency, and response time.

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#### 1. INTRODUCTION

The electric motor is the backbone of modern industry due to its ability to enhance process efficiency [1]–[3]. Electric motors are broadly classified based on their supply current into AC and DC types. AC motors are more favorable in terms of maintenance costs compared to DC motors, which require carbon brushes [4]. AC motors are further divided into synchronous and asynchronous types. The induction motor (IM), a subtype of asynchronous motor, is the most widely used due to its low cost, simple yet robust design, ease of maintenance, and high reliability [5][6]. In fact, IMs account for approximately 60% of a plant's total energy consumption [7]. They are also commonly used as traction motors in electric vehicles, such as those discussed in [8]–[11], and in railway applications [12].

Despite their widespread use, induction motors present challenges due to their complex and nonlinear mathematical behavior and lack of inherent speed control, making them traditionally suited for fixed-speed applications [13]. According to [14], the main limitation of IMs is their speed control capability. However, with advances in semiconductor technology, more sophisticated real-time control methods have become feasible.

One widely adopted technique for IM speed regulation is the Variable Frequency Drive (VFD). An overview of VFD control approaches is shown in Figure 1 [15]. VFDs are generally classified as either sensor-based or sensorless. The sensor-based category is further subdivided into scalar and vector control methods. Scalar control offers simple modeling and is parameter independent [16]. In contrast, vector control is more commonly used due to its superior performance [17]. Vector control includes two main techniques: field-oriented control (FOC) and direct torque control (DTC). FOC provides fast dynamic response and high efficiency, but is highly sensitive to parameter variations. DTC offers robust performance even under parameter fluctuations [18].

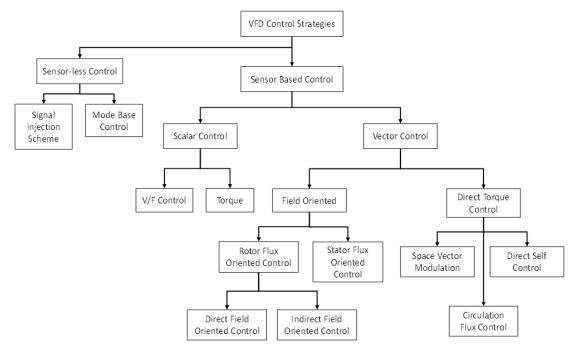


Figure 1. VFD control classification [15]

While FOC and DTC effectively manage torque, they do not directly handle speed regulation, necessitating additional control mechanisms. Numerous algorithms have been developed for IM speed control, including Proportional–Integral–Derivative (PID) [4],[19], Linear Quadratic Gaussian (LQG) [20], Sliding Mode Control (SMC) [21][22], Artificial Neural Networks (ANN) [23][24], Fuzzy Logic Controller (FLC) [25], and hybrid approaches that combine traditional and intelligent techniques [26]. Among these, PID control remains the most widely used due to its simplicity and robust general performance [27][28]. As noted in [29][30], PID is easily applicable across a wide range of industrial systems.

According to [31], nearly 90% of industries utilize PID controllers. The Ziegler–Nichols method is one of the most well-known PID tuning techniques. However, achieving optimal performance using this method can be difficult [32]. To overcome this, researchers have employed metaheuristic optimization algorithms such

as Ant Colony Optimization (ACO) [33], Genetic Algorithms (GA) [34], and Particle Swarm Optimization (PSO) [35] to fine-tune PID gains.

In this study, PSO is used to optimize PID parameters based on a defined objective function. PSO has been demonstrated to be effective in obtaining optimal control gains in various domains, as shown in [36][37]. Its fast convergence and ease of implementation have made it a popular choice for control tasks involving DC motors, robotic manipulators, and intelligent wheelchairs [38].

The main contributions of this paper are as follows:

- 1. It presents a simplified control design methodology by leveraging a small-signal state-space model derived from a detailed IM;
- 2. It demonstrates the effectiveness of PSO-tuned PID in improving steady-state performance using the Integral of Absolute Error (IAE) as the optimization criterion;
- 3. It compares the proposed method against a conventional PID controller under varying speed and load scenarios, showing significant error reduction without compromising transient stability.

This work aims to provide a computationally efficient yet effective tuning strategy for vector-controlled IMs, paving the way for future research into multi-objective optimization techniques that simultaneously consider robustness, energy efficiency, and transient performance. The remainder of this paper is organized as follows: Section II presents the IM modeling and the PSO algorithm used for PID tuning. Section III discusses the results and analysis. Section IV concludes the paper.

## 2. METHODS

# 2.1. System Model

The IM specifications used in this study are listed in Table 1. The detailed Simulink model of the IM system is illustrated in Figure 2, which includes both vector control and speed control blocks. This work focuses on optimizing the speed control component, while the vector control is based on the FOC method. However, performing PID optimization directly on the full detailed model (Figure 2) is computationally intensive and not practical for real-time or hardware applications. Therefore, a simplified state-space model is derived to enable efficient controller tuning. This reduced-order model provides a practical interface for offline PID tuning, which can later be implemented on real hardware systems.

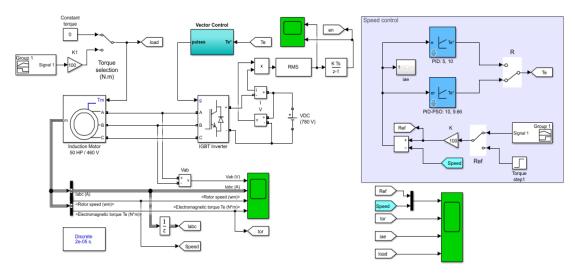


Figure 2. Induction motor detailed model in Simulink

Table 1. Induction motor parameters

Parameters	Values
Power	50 HP
Torque	300 Nm
Nominal Voltage	460 V
Nominal Frequency	60 Hz
Rotor type	Squirrel-cage

To approximate the system behavior, black-box modeling is employed using MATLAB's System Identification Toolbox. This approach extracts a state-space model from input-output data of the Simulink simulation. Unlike analytical or physics-based derivation (e.g., [39][40]), this method requires no prior

knowledge of motor parameters, making it suitable for real-world hardware where parameters may be unknown or difficult to measure [41]. The resulting state-space representation is given in (1), where x, u, and y denote the state, input, and output vectors, respectively:

$$\dot{x} = \begin{bmatrix} 1.378 & 0.285 \\ -1.25 & 0.06 \end{bmatrix} x + \begin{bmatrix} 0.16 \\ -0.52 \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} x$$
(1)

The modeling process is summarized in Figure 3. First, the combined system (FOC, inverter, and IM) is treated as a black box, with input as torque reference and output as motor speed. From this, a simplified linear model is obtained. This model is then used to derive a small-signal model, which allows for efficient offline PID tuning via PSO. Finally, the optimized PID parameters are validated in the detailed Simulink model and are intended for hardware implementation. The small-signal model neglects nonlinear dynamics such as rotor flux linkage variation, saturation effects, and parameter changes under load. Therefore, its accuracy in capturing transient behavior may be limited. Future work should include validation against the detailed model.

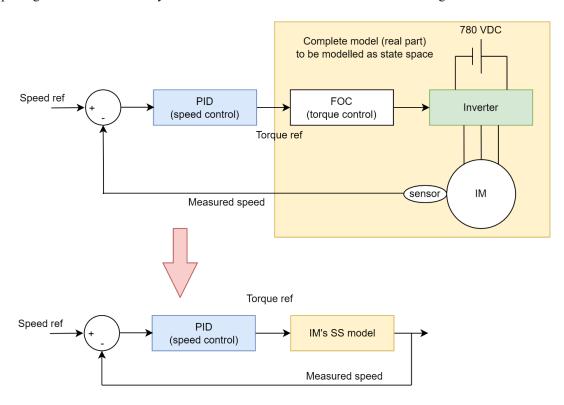


Figure 3. Modeling process from a detailed model (real hardware) into a small signal model

# 2.2. PID Tuned by PSO

PID control that must serve an objective function, such as optimal control, is referred to as optimal PID control. PSO is used to optimize PID parameters to satisfy the objective function's goal. Kennedy and Eberhart were the first to introduce PSO, which is based on the social and communicative behavior of birds [41]. Every individual is supposed to be a vector and offer a possible solution to optimization problems. In this study, the objective function used is IAE with the formula shown in (2). IAE is chosen over alternatives like ISE (Integral of Squared Error) or ITAE (Integral of Time-weighted Absolute Error) because it offers a balanced trade-off between responsiveness and robustness. It penalizes sustained errors without excessively amplifying brief transients, making it more suitable for speed control in induction motor systems [42]. At the same time, the parameters of the PSO are listed in Table 2.

$$IAE = \int_0^t |e(t)|dt \tag{2}$$

<b>Table 2.</b> PSO par	ameters
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Parameters	Values
Population	20
Max Iteration	10
Number of variables	2
Lower bound	0
Upper bound	10

The PID tuning process is illustrated in Figure 4. The algorithm starts by randomly initializing the PID parameters. Each candidate is tested using the small-signal model (from Figure 3) to evaluate its IAE. PSO then updates particles based on their own best performance and the global best result. This loop continues until the maximum number of iterations is reached. The best PID parameters are then applied to the detailed Simulink model (Figure 2) for validation and potential hardware implementation.

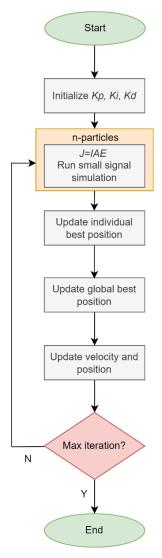


Figure 4. Flowchart of PSO tuning PID

# 3. RESULT AND DISCUSSION

To evaluate the performance of the proposed PID-PSO, two tests were conducted: a speed variation test and a load variation test. The system was designed with a 2% tolerance band for speed control. Performance was assessed using three key metrics: settling time, overshoot/undershoot (OS/US), and the IAE, which quantifies the accumulated tracking error throughout the simulation. Table 3 presents the manually tuned and PSO-optimized PID parameters used in the simulations.

Table 3. PID parameters			
Parameters	Manual	PSO	
$K_p$	5	10	
$K_i$	10	9.66	
$K_d$	0	0	

#### 3.1. Speed Variation Test

The motor was subjected to three step changes in speed: 50 rpm, 100 rpm, and 80 rpm. The corresponding simulation results for speed, torque, and IAE are shown in Figure 5. In Figure 5(a), both controllers successfully track the speed reference. Visually, PID-PSO shows a faster rise and smaller overshoot. However, quantitative analysis reveals that conventional PID achieves faster settling times for most steps when using a 2% tolerance criterion. This discrepancy arises because the PID response crosses the steady-state band earlier, even though it appears slower in the plot, highlighting the importance of defining performance metrics precisely. The torque responses in Figure 5(b) are similar between both methods, though PID-PSO produces sharper peaks during speed changes. The IAE comparison in Figure 5(c) shows that PID-PSO achieves a significantly lower cumulative error over the entire simulation. Quantitative performance data are summarized in Table 4. PID-PSO achieves a 37.79% reduction in cumulative IAE (from 52.26 to 34.08) compared to PID. Additionally, PID-PSO consistently reduces overshoot and undershoot, improving steady-state performance.

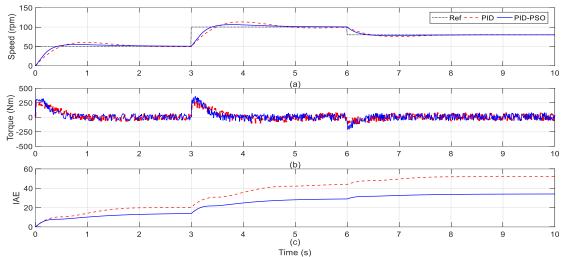


Figure 5. Speed variation test results

 Table 4. Results of speed variations

Control method	Settling time (s)	OS/ US (%)	IAE	
Set-point-1: 50rpm				
PID	1.98	20.64		
PID-PSO	2.29	9.40		
Set-point-2: 100rpm				
PID	1.75	13.32		
PID-PSO	1.83	6.28		
Set-point-3: 80rpm				
PID	1.66	-5.53	52.26	
PID-PSO	1.11	-2.29	34.08	

## 3.2. Load Variation Test

This test evaluated controller robustness under varying mechanical load conditions at a constant speed set-point of 100 rpm. The system was subjected to an increase and subsequent decrease in load. Figure 6(a) shows that PID-PSO produces lower overshoot and undershoot during load disturbances compared to PID. Torque responses in Figure 6(b) are similar for both controllers. As shown in Figure 6(c), PID-PSO again outperforms PID in terms of IAE, indicating better overall tracking performance. Table 5 summarizes the performance metrics. While the PID controller maintains a faster settling time (2.23 s vs. 2.93 s), PID-PSO achieves a 14.76% reduction in total IAE and significantly lower overshoot/undershoot during load transients.

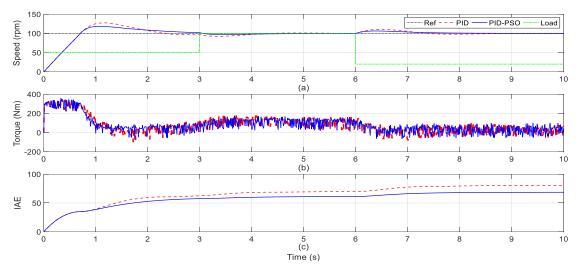


Figure 6. Load variation test results

Table 5. Results of load variations

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Control method	Settling time (s)	OS/ US (%)	IAE	
Set-point-1: 100rpm				
PID	2.23	27.51		
PID-PSO	2.93	18.53		
Increase load				
PID		7.58		
PID-PSO		3.13		
Decrease load				
PID		10.33	80.51	
PID-PSO		6.40	68.63	

The main findings confirm that the PID-PSO controller significantly reduces IAE and overshoot/undershoot across both test scenarios, although it introduces a modest increase in settling time compared to conventional PID. This trade-off between speed of response and accuracy is typical when optimizing a single objective (IAE), as done in this study. The limitation of this study is the optimization focused solely on IAE, overlooking other industrially relevant objectives such as robustness to parameter variation (e.g., rotor resistance changes) and actuator constraints (e.g., saturation). Despite these limitations, the study demonstrates the potential of PSO to improve control accuracy, especially in scenarios where overshoot and error accumulation are critical. Future work should consider multi-objective optimization and validate against more complex or real-time models.

#### 4. CONCLUSIONS

This study presented the design and simulation of an induction motor speed control system using a PID controller optimized by PSO (PID-PSO). A black-box modeling approach was employed to derive a reduced-order state-space model for PID tuning. The PSO algorithm was applied with the IAE as the objective function to enhance control accuracy. The controller's performance was evaluated through speed and load variation tests. While the manually tuned PID controller achieved faster settling times, the PID-PSO method significantly outperformed it in terms of overshoot, undershoot, and overall control accuracy. Specifically, PID-PSO reduced the IAE by 37.79% during speed variation and by 14.76% during load variation compared to manual PID tuning. These improvements indicate enhanced stability and tracking performance, which are critical for precise motor control. However, the modeling approach simplifies the nonlinear dynamics of the induction motor, which may limit real-world applicability. Future work should consider high-fidelity modeling and experimental validation to address this limitation. Moreover, integrating multi-objective optimization could further enhance the controller's ability to balance trade-offs between performance metrics such as settling time and transient behavior.

#### **DECLARATION**

# **Author Contribution**

All authors contributed equally to the main contributor to this paper. All authors read and approved the final paper.

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#### Conflicts of Interest

The authors declare no conflict of interest.

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